

**Solve the following questions:-**

**Question One** (20 Mark)

1. Determine the output voltage for the circuit in Fig. 1,

(6 marks)

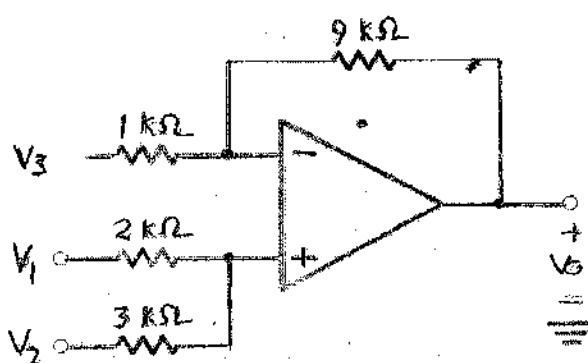


Fig . 1

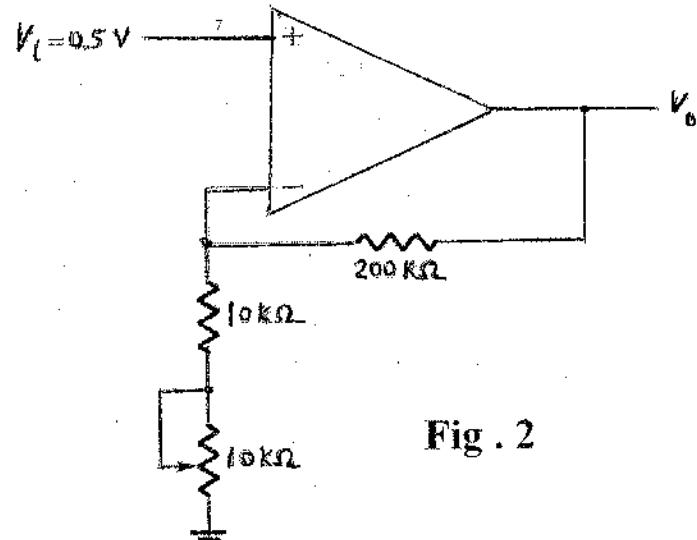


Fig . 2

2. What range of output voltage is developed in the circuit of Fig. 2

(7 marks)

3. Define CMRR and how to experimentally measure CMRR?

(7 marks)

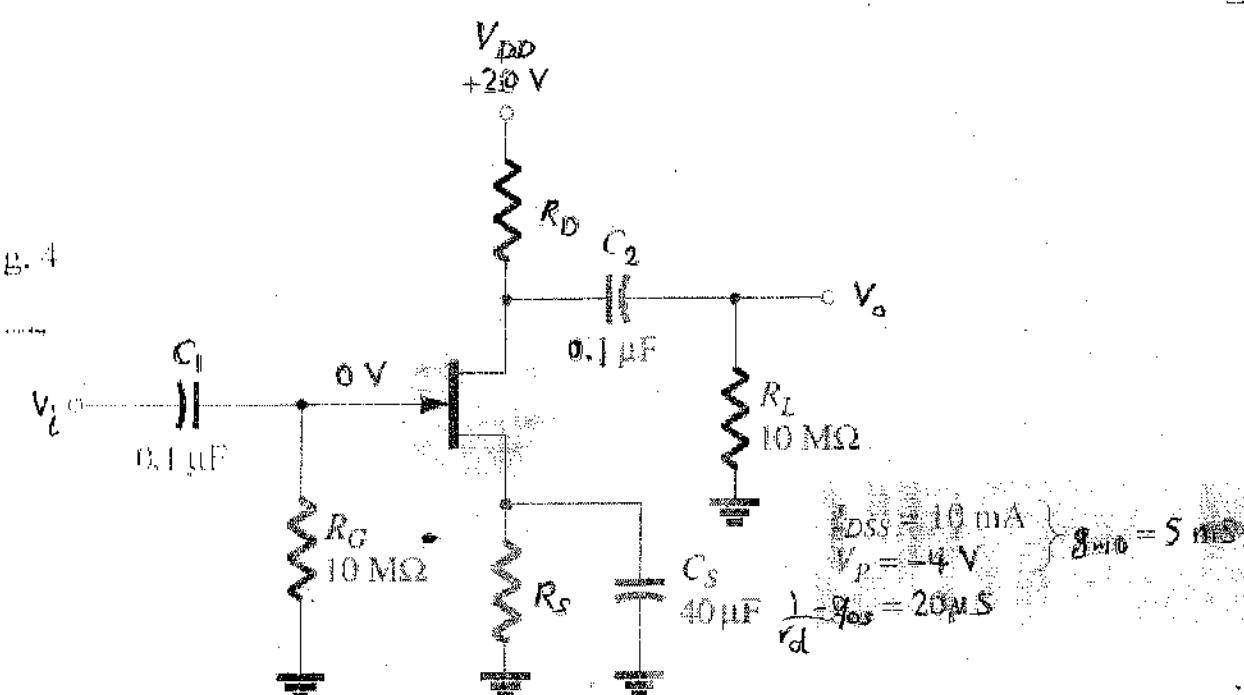
**Question Two** (20 Mark)

1. Calculate the efficiency of a class B amplifier for a supply voltage of  $V_{CC} = 22$  V driving a  $4 \Omega$  load with peak output voltages of: a.  $V_L(p) = 20$  V. b.  $V_L(p) = 4$  V.
2. A transformer-coupled class A amplifier drives a  $16 \Omega$  speaker through a 3.87:1 transformer. Using a power supply of  $V_{CC} = 36$  V, the circuit delivers 2 W to the load. Calculate:
  - a. P(ac) across transformer primary.
  - b.  $V_L(ac)$ .
  - c.  $V(ac)$  at transformer primary.
  - d. The rms values of load and primary current.
3. Explain briefly three different configurations of practical class B power amplifier

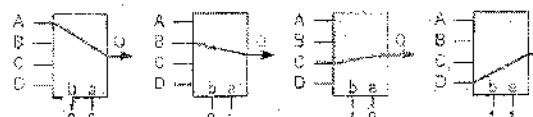
**Question Three** (30 Mark)

- 1- Choose the values of  $R_D$  and  $R_S$  for the network of Fig. 3 that will result in a gain of 8 using a relatively high level of  $g_m$  for this device defined at  $V_{GSQ} = 0.25V_P$ .

Fig. 4



- 2- An analog switch uses an n-channel MOSFET with  $V_{GS(th)} = 4$  V. A voltage of +8 V is applied to the gate. Determine the maximum peak-to-peak input signal that can be applied if the drain-to-source voltage drop is neglected.
- 3- An analog switch is used to sample a signal with  $x(t) = 10 + 5\cos 2000t + 8 \cos 8000t$ . Determine the minimum frequency of the pulses applied to the MOSFET gate.
- 4- Design a quad time division multiplexer using an n-channel MOSFET with  $V_{GS(th)} = 3$  V. A voltage of +8 V is applied to the gate of each MOSFET during 0.25 μs. Calculate the sampler frequency and maximum signal frequency.



Best wishes of success

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$$\begin{aligned} V &= 2 \times \frac{V_2}{5} \quad (1+4) = 4 \sqrt{2} \\ &\Rightarrow \frac{3V_1}{5} (1+4) = 6 \sqrt{2} \end{aligned}$$

## Model answer

### Q1-1:

If in the circuit of Fig. E2.9 the 1-k $\Omega$  resistor is disconnected from ground and connected to a third signal source  $v_3$ , use superposition to determine  $v_o$  in terms of  $v_1$ ,  $v_2$ , and  $v_3$ .

Ans.  $v_o = \cancel{6V} + 4v_1 - 9v_2 - 12v_3$

### Q1-2:

$$V_o = \left(1 + \frac{R_2}{R_1}\right) V_1$$

For  $R_1 = 10\text{ k}\Omega$ :

$$V_o = \left(1 + \frac{200\text{ k}\Omega}{10\text{ k}\Omega}\right)(0.5\text{ V}) = 21(0.5\text{ V}) = 10.5\text{ V}$$

For  $R_1 = 20\text{ k}\Omega$ :

$$V_o = \left(1 + \frac{200\text{ k}\Omega}{20\text{ k}\Omega}\right)(0.5\text{ V}) = 11(0.5\text{ V}) = 5.5\text{ V}$$

$V_o$  ranges from 5.5 V to 10.5 V.

### Q1-3: Common-Mode Rejection Ratio

The ratio between  $A_d$  and  $A_c$  is the common-mode rejection ratio (CMRR),

$$\text{CMRR} = A_d / A_c$$

1. To measure  $A_d$ : Set  $V_{i1} = -V_{i2} = V_s = 0.5\text{ V}$ , so that

$$V_d = (V_{i1} - V_{i2}) = (0.5\text{ V} - (-0.5\text{ V})) = 1\text{ V}$$

and

$$V_c = 1/2(V_{i1} + V_{i2}) = 1/2[0.5\text{ V} + (-0.5\text{ V})] = 0\text{ V}$$

Under these conditions the output voltage is

$$V_o = A_d V_d + A_c V_c = A_d (1\text{ V}) + A_c (0) = A_d$$

Thus, setting the input voltages  $V_{i1} = -V_{i2} = 0.5\text{ V}$  results in an output voltage numerically equal to the value of  $A_d$ .

2. To measure  $A_c$ : Set  $V_{i1} = V_{i2} = V_s = 1\text{ V}$ , so that

$$V_d = (V_{i1} - V_{i2}) = (1\text{ V} - 1\text{ V}) = 0\text{ V}$$

$$\text{and } V_c = 1/2(V_{i1} + V_{i2}) = 1/2(1\text{ V} + 1\text{ V}) = 1\text{ V}$$

Under these conditions the output voltage is

$$V_o = A_d V_d + A_c V_c = A_d (0\text{ V}) + A_c (1\text{ V}) = A_c$$

Thus, setting the input voltages  $V_{i1} = V_{i2} = 1\text{ V}$  results in an output voltage numerically equal to the value of  $A_c$ .

$$\text{CMRR} = A_d / A_c$$

Q2-1:

(a)  $V_{L_{(\text{peak})}} = 20 \text{ V}$

$$P_i = V_{CC}I_{dc} = V_{CC} \left[ \frac{2}{\pi} \cdot \frac{V_L}{R_L} \right]$$
$$= (22 \text{ V}) \left[ \frac{2}{\pi} \cdot \frac{20 \text{ V}}{4 \Omega} \right] = 70 \text{ W}$$

$$P_o = \frac{V_L^2}{2R_L} = \frac{(20 \text{ V})^2}{2(4 \Omega)} = 50 \text{ W}$$

$$\% \eta = \frac{P_o}{P_i} \times 100\% = \frac{50 \text{ W}}{70 \text{ W}} \times 100\% = 71.4\%$$

(b)  $P_i = (22 \text{ V}) \left[ \frac{2}{\pi} \cdot \frac{4 \text{ V}}{4 \Omega} \right] = 14 \text{ W}$

$$P_o = \frac{(4)^2}{2(4)} = 2 \text{ W}$$

$$\% \eta = \frac{P_o}{P_i} \times 100\% = \frac{2 \text{ W}}{14 \text{ W}} \times 100\% = 14.3\%$$

Q2-2

(a)  $P_{pri} = P_L = 2 \text{ W}$

(b)  $P_L = \frac{V_L^2}{R_L}$

$$V_L = \sqrt{P_L R_L} = \sqrt{(2 \text{ W})(16 \Omega)}$$
$$= \sqrt{32} = 5.66 \text{ V}$$

$$(c) \quad R_2 = a^2 R_1 = (3.87)^2 (16 \Omega) = 239.6 \Omega$$

$$P_{pri} = \frac{V_{pri}^2}{R_{pri}} = 2 \text{ W}$$

$$V_{pri}^2 = (2 \text{ W})(239.6 \Omega)$$

$$V_{pri} = \sqrt{479.2} = 21.89 \text{ V}$$

$$[\text{or, } V_{pri} = aV_L = (3.87)(5.66 \text{ V}) = 21.9 \text{ V}]$$

$$(d) \quad P_L = I_L^2 R_L$$

$$I_L = \sqrt{\frac{P_L}{R_L}} = \sqrt{\frac{2 \text{ W}}{16 \Omega}} = 353.55 \text{ mA}$$

$$P_{pri} = 2 \text{ W} = I_{pri}^2 R_{pri} = (239.6 \Omega) I_{pri}^2$$

$$I_{pri} = \sqrt{\frac{2 \text{ W}}{239.6 \Omega}} = 91.36 \text{ mA}$$

$$\text{or, } I_{pri} = \frac{I_L}{a} = \frac{353.55 \text{ mA}}{3.87} = 91.36 \text{ mA}$$

**Solution:** The operating point is defined by

$$V_{GSQ} = \frac{1}{4}V_P = \frac{1}{4}(-4\text{ V}) = -1\text{ V}$$

and  $I_D = I_{DSS} \left(1 - \frac{V_{GSQ}}{V_P}\right)^2 = 10\text{ mA} \left(1 - \frac{(-1\text{ V})}{(-4\text{ V})}\right)^2 = 5.625\text{ mA}$

Determining  $g_m$ , we obtain

$$\begin{aligned} g_m &= g_{m0} \left(1 - \frac{V_{GSQ}}{V_P}\right) \\ &= 5\text{ mS} \left(1 - \frac{(-1\text{ V})}{(-4\text{ V})}\right) = 3.75\text{ mS} \end{aligned}$$

The magnitude of the ac voltage gain is determined by

$$|A_v| = g_m(R_D \parallel r_d)$$

Substituting known values results in

$$8 = (3.75\text{ mS})(R_D \parallel r_d)$$

so that

$$R_D \parallel r_d = \frac{8}{3.75\text{ mS}} = 2.13\text{ k}\Omega$$

The level of  $r_d$  is defined by

$$r_d = \frac{1}{g_{m0}} = \frac{1}{20\text{ }\mu\text{S}} = 50\text{ k}\Omega$$

and

$$R_D \parallel 50\text{ k}\Omega = 2.13\text{ k}\Omega$$

with the result that

$$R_D = 2.2\text{ k}\Omega$$

which is a standard value.

The level of  $R_S$  is determined by the dc operating conditions as follows:

$$V_{GSQ} = -I_D R_S$$

$$-1\text{ V} = -(5.625\text{ mA}) R_S$$

and

$$R_S = \frac{1\text{ V}}{5.625\text{ mA}} = 177.8\text{ }\Omega$$

The closest standard value is  $180\text{ }\Omega$ . In this example,  $R_S$  does not appear in the ac design because of the shorting effect of  $C_S$ .

$$\begin{aligned} P_3-2.0 \quad V_{GP-P} &= 3\text{ V} \\ P_3-3 &= \frac{\text{f}_{\text{sample}}}{\text{f}_{\text{min}}} \geq 2 \text{ f}_{\text{signal max}} \\ &\geq \frac{3000}{2\pi} \text{ Hz} \\ \text{f}_{\text{min}} &\geq \frac{2\pi}{3000} \text{ Hz} = 2.54\text{ kHz} \end{aligned}$$

$$\begin{array}{c} D \\ \hline S \\ \hline T \\ \hline G = V_P \end{array}$$

$$\begin{aligned} V_{GS} &= V_G - V_S \geq 1\text{ V} \\ V_S &= V_D = V_P \\ V_G &= 8\text{ V} \end{aligned}$$

$$\begin{aligned} P_3-4 \quad f_{\text{sample}} &= \frac{1}{2\pi R_S} = 4\text{ MHz} \\ f_{\text{signal max}} &= \frac{4\text{ MHz}}{4} = 1\text{ MHz} \end{aligned}$$

$$\begin{aligned} 8 - V_P &= 4\text{ V} \\ V_P - V_D &\leq 8\text{ V} \end{aligned}$$