

نموذج الاجابة الخاص بامتحان لثمن الآلى

د/ حرس طبري

Apply K.V.L.

$$V(t) = R i(t) + L \frac{di(t)}{dt} + E_b(t)$$

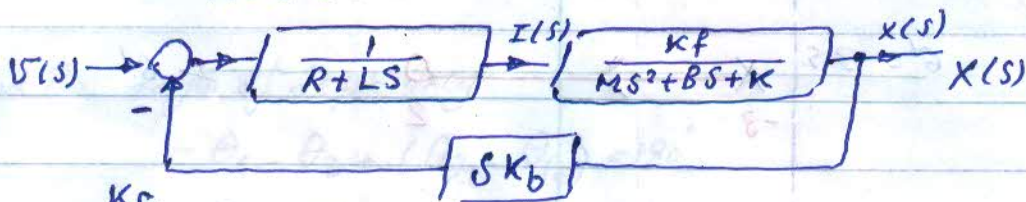
$$I(s) = \frac{V(s) - E_b(s)}{R + Ls} \quad (1)$$

$$E_b \propto \dot{x}(t) \quad \text{so, } E_b = K_b \dot{x}(t)$$

$$E_b(s) = K_b s X(s) \quad (2)$$

$$M \ddot{x}(t) = f(t) - Kx(t) + B \dot{x}(t) \quad f(t) \rightarrow \boxed{M} \leftarrow \begin{matrix} Kx(t) \\ B\dot{x}(t) \end{matrix}$$

$$X(s) = \left(\frac{K_f}{Ms^2 + Bs + K} \right) I(s) \quad (3)$$



$$\frac{X(s)}{V(s)} = \frac{K_f}{LMS^2 + (LB + RM)S^2 + (KL + RB + K_b K_f)S + RK}$$

Q1.1

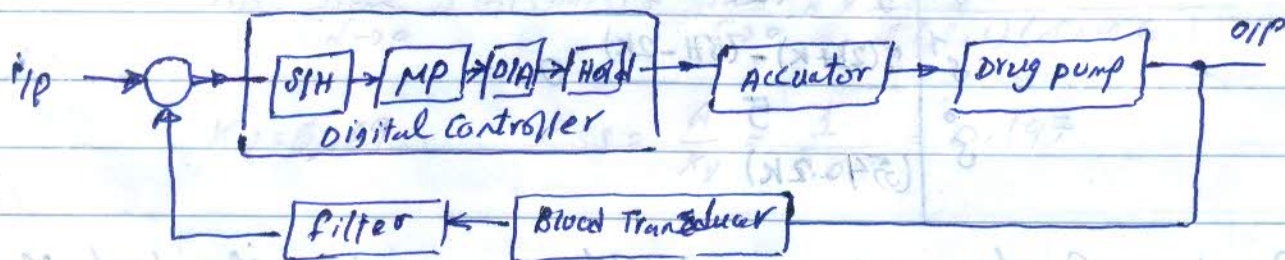
Set $b_1 = 1$ $V_F = 5(2^{-1}) = 2.5$, $U_x > 2.5$ leave $b_1 = 1$

Set $b_2 = 1$ $V_F = 2.5 + 5(2^{-2}) = 3.75$, $U_x < 3.75$ reset $b_2 = 0$

Set $b_3 = 1$ $V_F = 2.5 + 5(2^{-3}) = 3.125$, $U_x > 3.125$ leave $b_3 = 1$

Set $b_4 = 1$ $V_F = 3.125 + 5(2^{-4}) = 3.4375$, $U_x < 3.4375$ reset $b_4 = 0$

By this procedure, we find the output is a binary word of 1010.

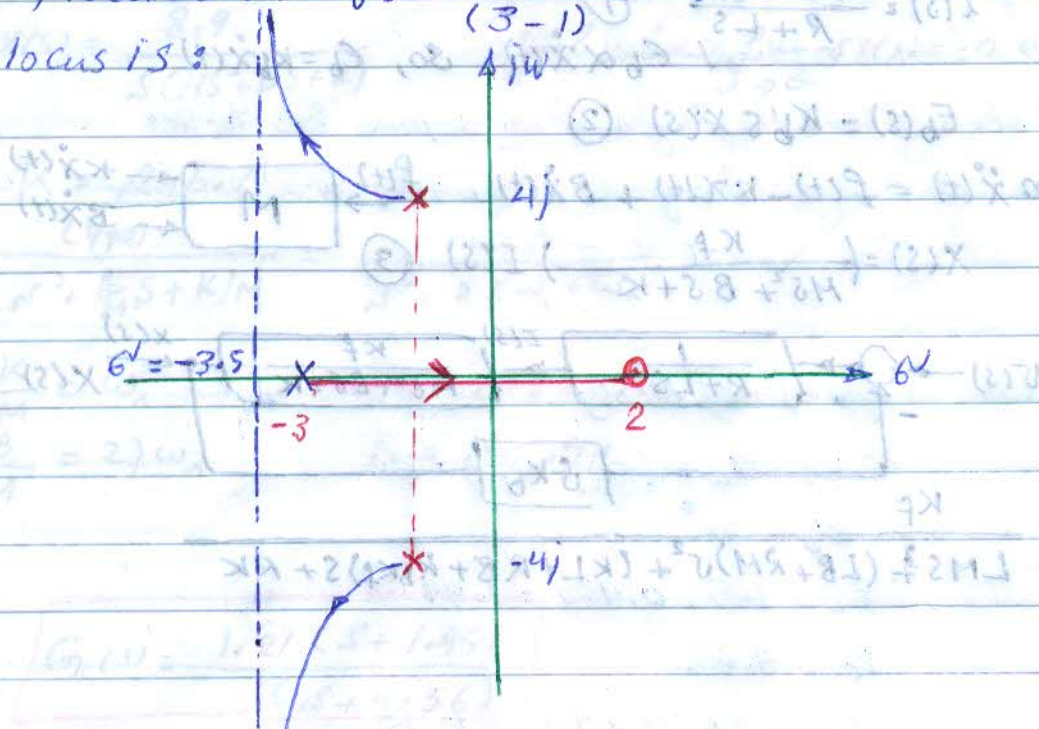


(Q.1.3)

(a) The pole and zeros of $P(s)$ are $P_1 = -3$, $P_{2,3} = -1 \pm j4$ & $Z_1 = 2$
The relative degree is $n - m = 2$. Therefore there exist two asymptotes

at $\pm 90^\circ$, located at $\sigma = \frac{(-3 - 1 - 1) - 2}{(3 - 1)} = -3.5$

The root locus is:



b) The characteristic equation is

$$(s + 3)(s^2 + 2s + 17) + k(s - 2) = 0$$

$$s^3 + 5s^2 + (23 + k)s + (51 - 2k) = 0$$

s^3	1	$(23 + k)$
s^2	5	$(51 - 2k)$
s^1	$5(23 + k) - (51 - 2k)$	0
s^0	5	
s	$(51 - 2k)$	

Based on Routh's criterion, stability is guaranteed if and only if

$$51 - 2k > 0 \quad \text{i.e.} \quad k < 25.5$$

Therefore: $0 < k < 25.5$

Q2.1

$$M_p = e^{-\pi \zeta / \sqrt{1-\zeta^2}}$$

$$\sin(0.15) = \frac{-\pi \zeta}{\sqrt{1-\zeta^2}} \quad \zeta = 0.5$$

$$T_d = \frac{1-0.7\zeta}{\omega_n} = 0.435 \quad \omega_n = 3 \text{ rad/sec}$$

$$s_0 = -\zeta \omega_n \pm j \omega_n \sqrt{1-\zeta^2}$$
$$= -1.5 \pm j 2.598$$

It's obvious that the design using a pure P-controller does not lead to the goal as the root locus does not traverse the dominant point

Find the angle of deficiency

$$-\theta_1 - \theta_2 + (\theta_{z_c} - \theta_{p_c}) = 180^\circ$$

$$\theta_D = 41^\circ$$

$$z_c = 1.5 + 0.435 = 1.94$$

$$p_c = 4.65$$

$$K = \frac{3 \times 2.7 \times 4.2}{2 \times 8} = 12.15$$

$$G_c = \frac{1.215(s+1.94)}{(s+4.65)}$$

$$G_c G_p = \frac{12.5(s+1.94)}{s(s+1)(s+4.65)}$$

$$K_V = \lim_{s \rightarrow 0} s G(s) H(s) = \lim_{s \rightarrow 0} s \times \frac{12.5(s+1.94)}{s(s+1)(s+4.65)}$$

$$K_V = 5.069 \quad e_{ss} = \frac{A}{K_V} = \frac{1}{5.069} = 0.197$$

Q2.1

$$G_p = \frac{K_p \pi (s+z_i)}{s' \pi (s+p_i)}, \quad G_c = \frac{K_c (s+z_c)}{(s+p_c)} \quad G_p G_c = \frac{K K_c (s+z_i)(s+z_c)}{s' (s+p_c) \pi (s+p_i)}$$

let $J=1$

$$K_V = \lim_{s \rightarrow 0} s G_p G_c = K \frac{z_c}{p_c} \cdot \frac{\pi z_i}{\pi p_i} ; \text{ The compensator is lead } p_c > z_c$$

$\therefore \frac{z_c}{p_c} < 1 \quad K_V \text{ will decrease}$

$$e_{ss} = \frac{A}{K_V} \text{ Steady state error will increase.}$$